

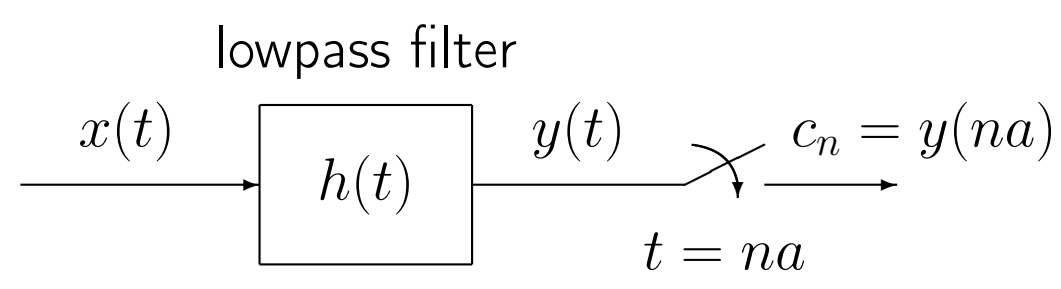
Motivation – Causal Reconstruction

Causal signal reconstruction is crucial

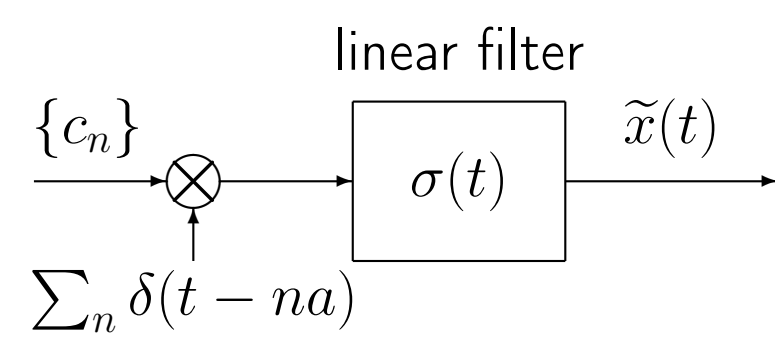
- in on-line applications
- feedback loops in control systems
- to reduce border effects in image processing, etc.

Framework – Shift Invariant Sampling

Non-ideal Acquisition



Shift-Invariant Reconstruction



$$c_n = y(na) = \int_{-\infty}^{\infty} h(\tau) x(na - \tau) d\tau$$

$$\tilde{x}(t) = \sum_{n \in \mathbb{Z}} c_n \sigma(t - na)$$

- Subordinate signal space: arbitrary Hilbert space \mathcal{H}
- The sampling process can be described as the evaluation inner products

$$c_n = y(na) = \langle x, s_n \rangle, \quad n \in \mathbb{Z}$$

- Sampling functions $s_n \in \mathcal{H}$ have the form $s_n = (T_a^n s)(t) = s(t - na)$

or in general $s_n = U^n s$ with $\begin{cases} \text{generator } s \in \mathcal{H} \\ U \text{ unitary operator on } \mathcal{H} \end{cases}$

⇒ Sequence $\mathbf{s} = \{s_n\}_{n \in \mathbb{Z}}$ of sampling functions forms a stationary sequence in \mathcal{H}

⇒ Sequence \mathbf{s} is characterized by its corresponding spectral density $\Phi_s \in L^1(\mathbb{T})$

$$\langle s_n, s_m \rangle = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{i(n-m)\theta} \Phi_s(e^{i\theta}) d\theta$$

- Sampling space: $\mathcal{S} := \overline{\text{span}}\{s_n : n \in \mathbb{Z}\}$

- $\mathbf{s} = \{s_n\}_{n \in \mathbb{Z}}$ is a Riesz basis for the sampling space \mathcal{S}

$$\Leftrightarrow 0 < A \leq \Phi_s(e^{i\theta}) \leq B < \infty \text{ for a.e. } \theta \in [-\pi, \pi]$$

■ M. Unser and A. Aldroubi, "A General Sampling Theory for Nonideal Acquisition Devices," *IEEE Trans. Signal Process.*, vol. 42, no. 11, pp. 2915–2925, Nov. 1994.

■ T. Michaeli, V. Pohl, and Y. C. Eldar, "U-Invariant Sampling: Extrapolation and Causal Interpolation from Generalized Samples," *IEEE Trans. Signal Process.*, vol. 59, no. 5, pp. 2085–2100, May 2011.

Reconstruction – Ideal World

Assumptions

- Let $x \in \mathcal{S}$ be an arbitrary signal
- All past and future signal samples $c_n = \langle x, s_n \rangle$, $n \in \mathbb{Z}$ are known

Goal

Signal reconstruction of the form

$$\tilde{x}(t) = \sum_{n \in \mathbb{Z}} \langle x, s_n \rangle \sigma_n(t)$$

such that

- $\tilde{x}(t) = x(t)$ for all $x \in \mathcal{S}$ (Perfect reconstruction)
- $\langle \tilde{x}, s_n \rangle = \langle x, s_n \rangle$ for all $n \in \mathbb{Z}$ (Consistency)

Solution

A well known result from frame theory states that the problem is solved by the dual Riesz basis $\{\sigma_n\}_{n \in \mathbb{Z}}$ of $\{s_n\}_{n \in \mathbb{Z}}$, given by

$$\sigma_n(t) = (U^n \sigma)(t) \quad (1)$$

$$\text{with } \sigma(t) = \sum_{k \in \mathbb{Z}} \alpha_k s_{-k}(t) \text{ and } \alpha_k = \frac{1}{2\pi} \int_{-\pi}^{\pi} \frac{e^{-ik\theta}}{\Phi_s(e^{i\theta})} d\theta.$$

Often, in reality only the past signal samples are known!

Reconstruction – Real World

New Assumption

Let $\mathbf{c}_0 = \{c_0, c_{-1}, \dots\}$ be the past signal samples known at $t = t_0$

Goal

Reconstruct the past signal component

$$x_-(t) = \begin{cases} x(t) & \text{if } t < t_0 \\ 0 & \text{if } t \geq t_0 \end{cases}$$

of the signal $x(t)$ at time $t = t_0$ from the past signal samples \mathbf{c}_0 only.

Naive Solution

$$\tilde{x}_-(t) = \sum_{n=0}^{\infty} c_{-n} \sigma_{-n}(t), \quad t \leq t_0$$

based on the non-causal dual frame (1).

However this reconstruction is not perfect, i.e. $\tilde{x}_-(t) \neq x_-(t)$, because we need the dual Riesz basis $\{\zeta_{-n}\}_{n=0}^{\infty}$ of $\{s_{-n}\}_{n=0}^{\infty}$!

Main Result – Causal Dual Riesz Basis

Theorem Let $\mathbf{s} = \{s_n\}_{n \in \mathbb{Z}}$ be a stationary sequence in a Hilbert space \mathcal{H} which is a Riesz basis for $\mathcal{S} = \overline{\text{span}}\{s_n : n \in \mathbb{Z}\}$ and let Φ_s be the spectral density of \mathbf{s} . Then $\mathbf{s}_0 = \{s_{-n}\}_{n=0}^{\infty}$ is a Riesz basis for $\mathcal{S}_0 = \overline{\text{span}}\{s_{-n} : n = 0, 1, 2, \dots\}$ and the corresponding dual Riesz basis $\{\zeta_{-n}\}_{n=0}^{\infty}$ is given by

$$\zeta_{-n} = \sum_{k=0}^{\infty} \hat{\psi}_n(k) s_{-k}, \quad n = 0, 1, 2, \dots$$

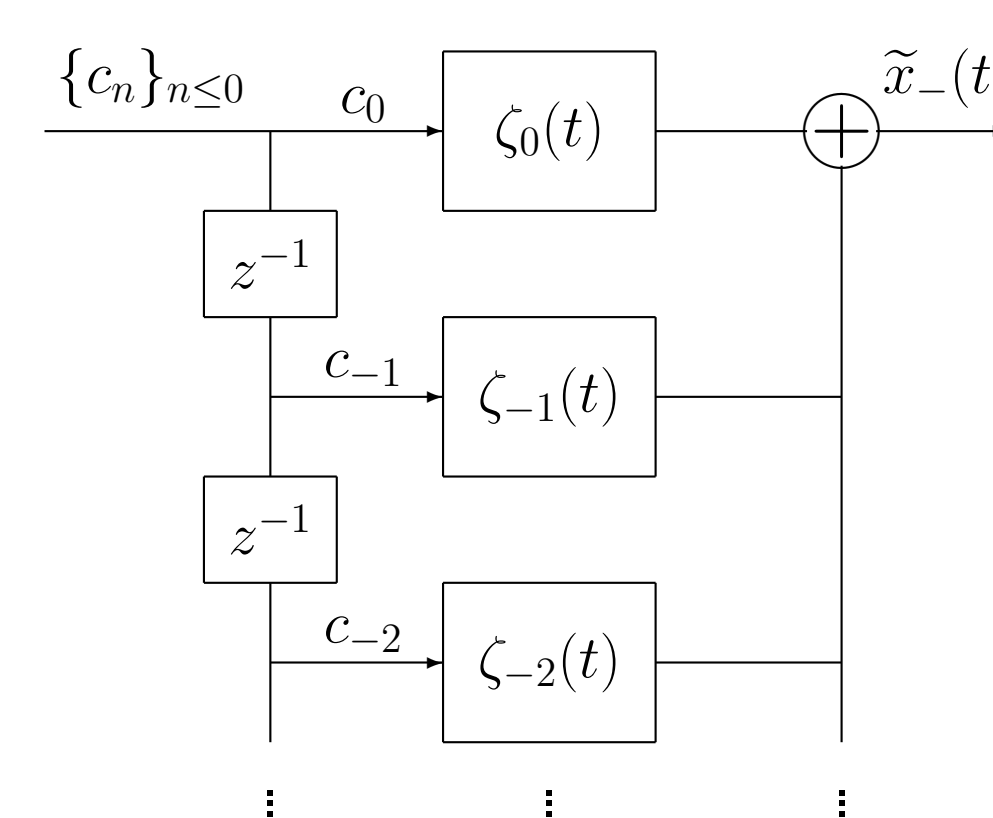
$$\text{with } \hat{\psi}_n(k) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \psi_n(e^{i\theta}) e^{-ik\theta} d\theta, \quad n = 0, 1, 2, \dots$$

$\psi_n \in H^2$ are defined as

$$\psi_n(e^{i\theta}) = \frac{1}{\Phi_s^+(e^{i\theta})} P_+ \left[\frac{e^{in\theta}}{\Phi_s^-(e^{i\theta})} \right], \quad n = 0, 1, 2, \dots$$

and wherein Φ_s^+ and Φ_s^- are the spectral factors of Φ_s .

Overall Causal Reconstruction Scheme



- Reconstructed past signal

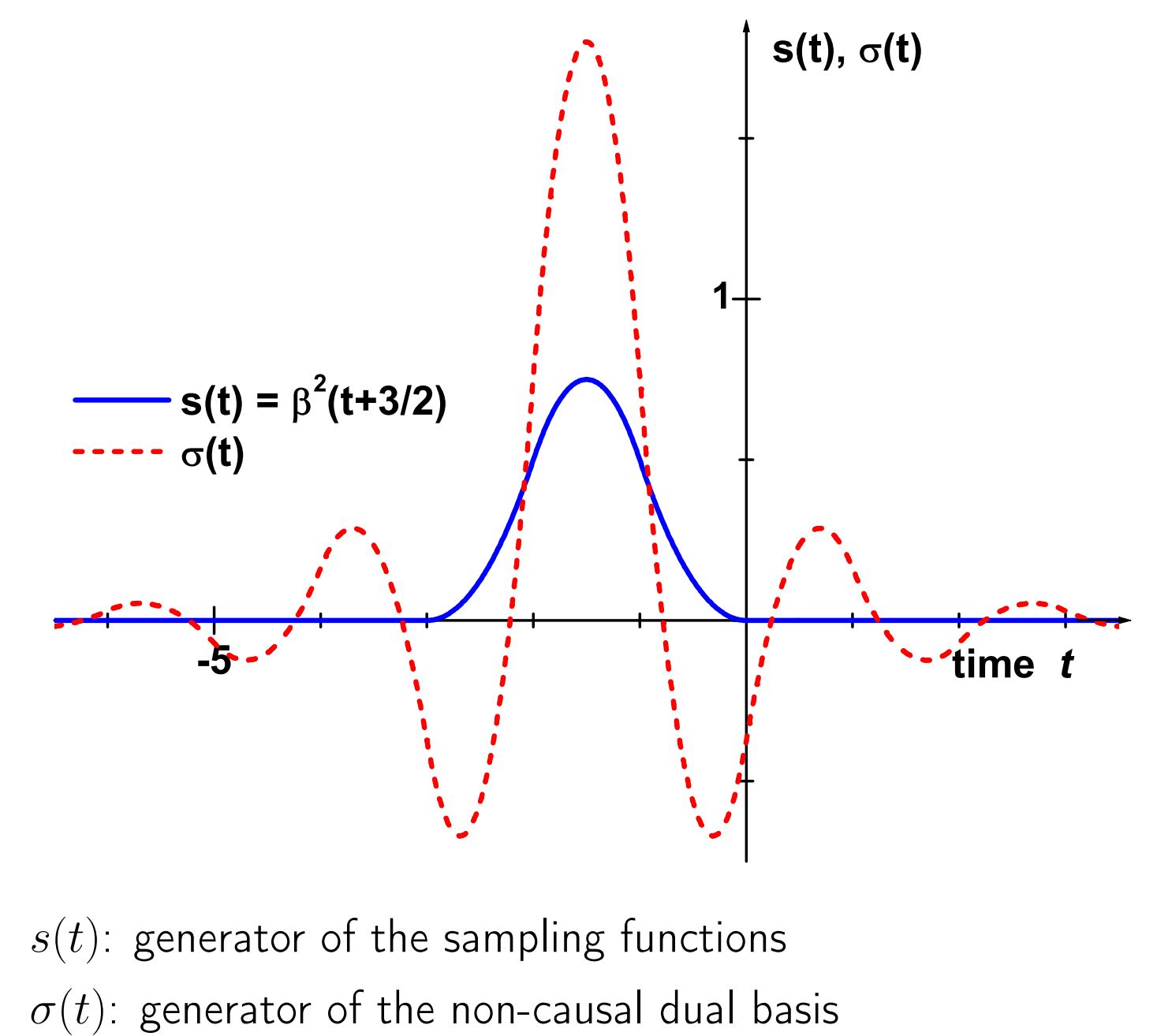
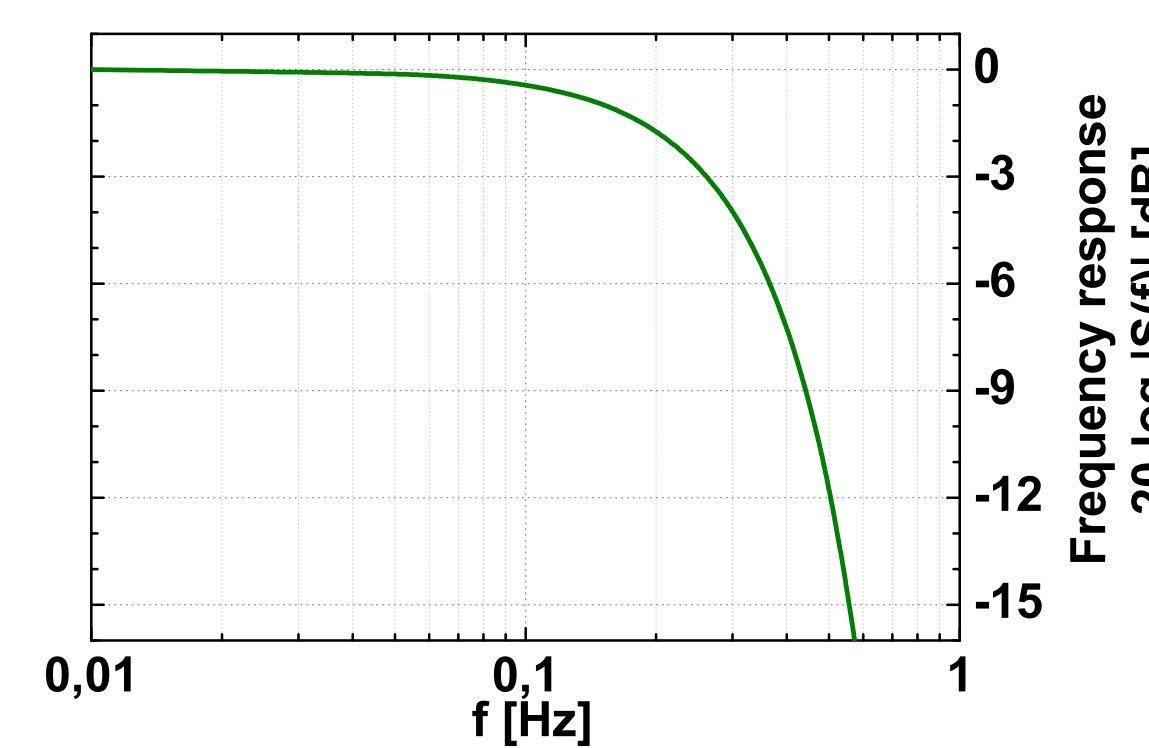
$$\tilde{x}_-(t) = \sum_{n \leq 0} c_n \zeta_{-n}(t), \quad t \leq t_0$$

- Each reconstruction kernel $\zeta_{-n}(t)$ has a different shape
- Each sampling value corresponds to the weighting of the respective kernel

Example – Causal Spline Reconstruction

Practical Assumptions

- Shift-invariant sampling in $L^2(\mathbb{R})$ with period $a = 1$: $(U^n s)(t) = s(t - n)$
- Impulse response $s(t)$ is a B-spline of 2nd degree as a model of a non-ideal lowpass



Causal versus Non-causal Reconstruction Kernels

- Dashed lines: truncated non-causal reconstruction kernels

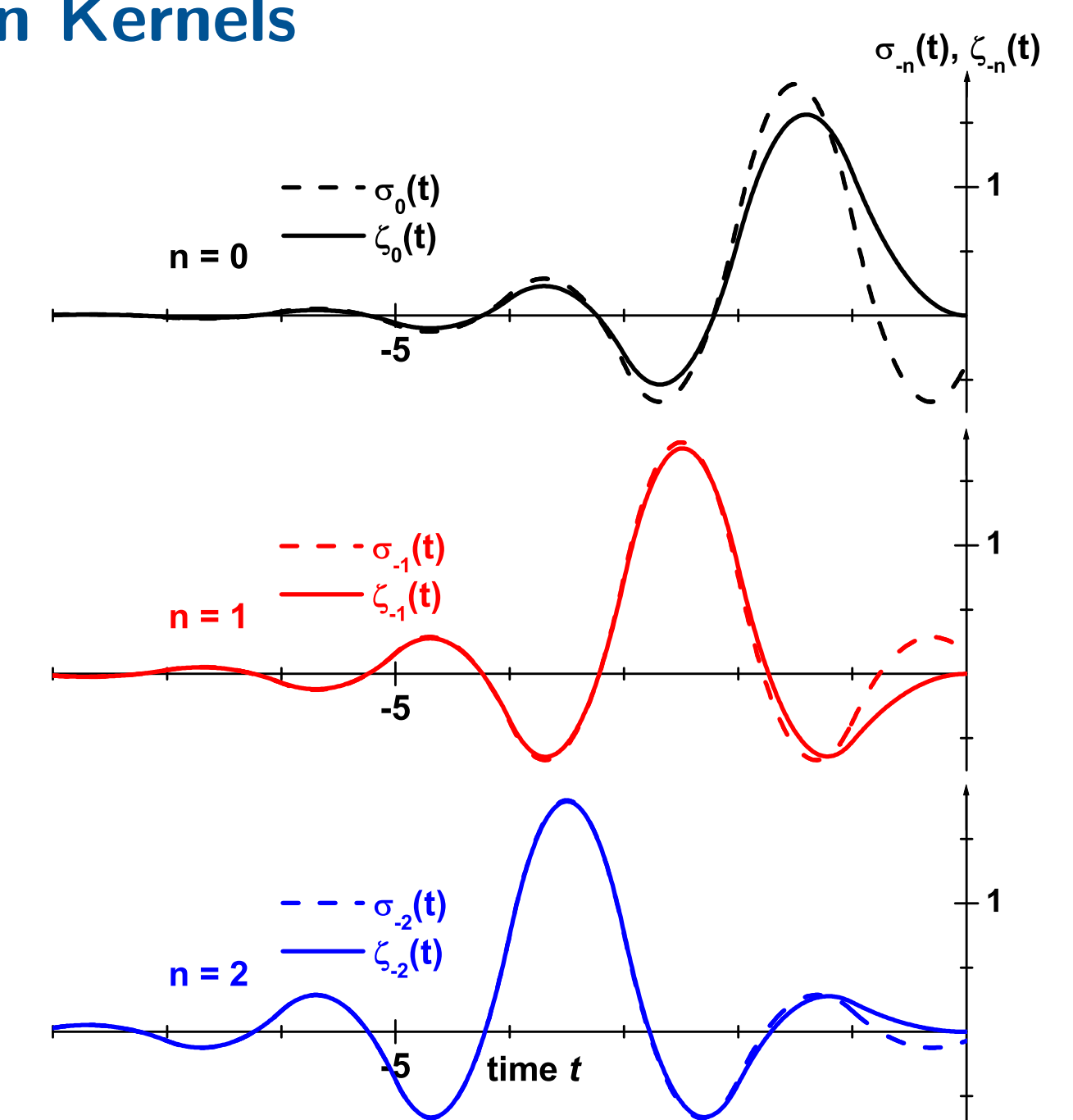
$$\sigma_{-n}(t) = \sum_{k \in \mathbb{Z}} \alpha_k s_{-k}(t + n)$$

- Solid lines: causal reconstruction kernels

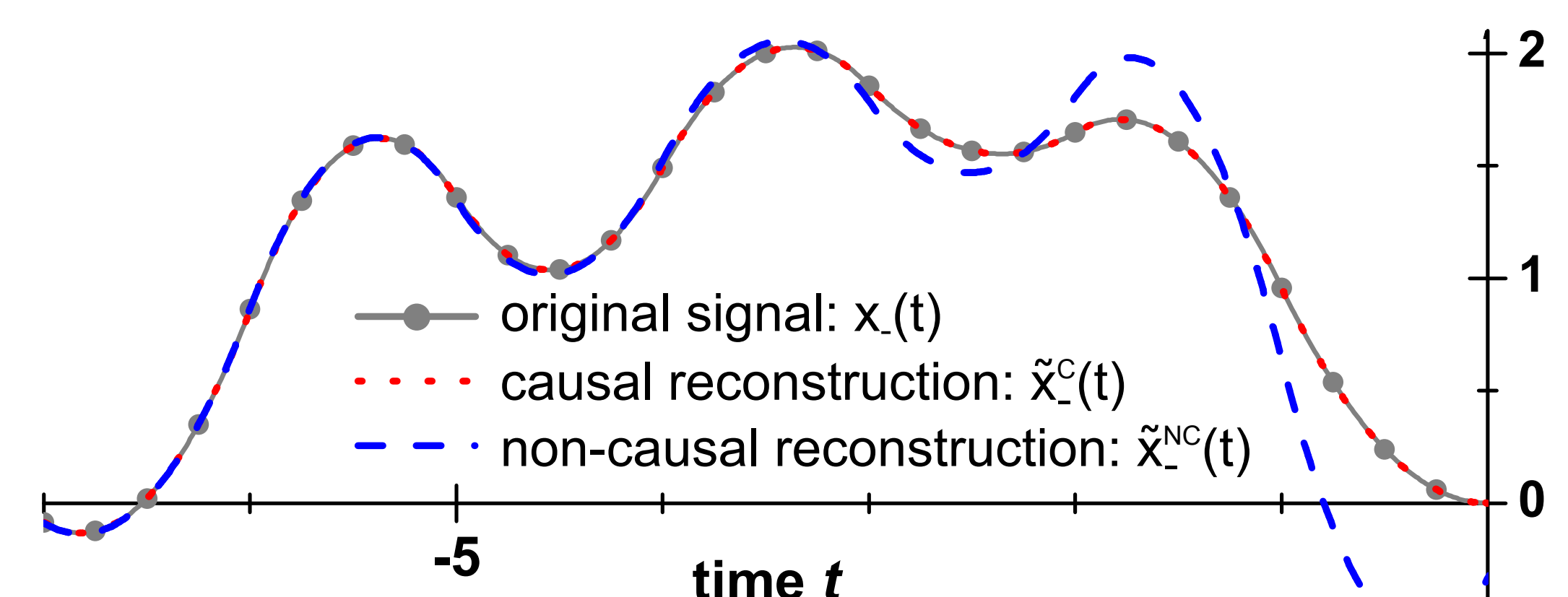
$$\zeta_{-n}(t) = \sum_{k=0}^{\infty} \hat{\psi}_n(k) s_{-k}(t)$$

- For large n , the causal kernel $\zeta_{-n}(t)$ converges to non-causal kernel $\sigma_{-n}(t)$

$$\zeta_{-n} \rightarrow \sigma_{-n} \text{ as } n \rightarrow \infty$$



Signal Reconstruction – Causal versus Non-causal



$$\tilde{x}_-^{\text{NC}}(t) = \sum_{n=0}^{\infty} \langle x, s_{-n} \rangle \sigma_{-n}(t) \quad \text{and} \quad \tilde{x}_-^{\text{C}}(t) = \sum_{n=0}^{\infty} \langle x, s_{-n} \rangle \zeta_{-n}(t)$$

- Significant differences close to the border
- Causal and non-causal reconstruction coincide at the distant past $t \rightarrow -\infty$